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# **Research Paper**

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Email: [xycaoBL@163.com](mailto:xycaoBL@163.com)

# An acceleration method employing sparse sensing matrix for fast analysis of the wide-angle electromagnetic problems based on compressive sensing

Qi Qi<sup>1</sup> (D[,](https://orcid.org/0000-0003-0739-7031) Xinyuan Cao<sup>1</sup>, Yi Liu<sup>1,2</sup>, Meng Kong<sup>1</sup>, Xiaojing Kuang<sup>1</sup> and Mingsheng Chen<sup>1</sup>

<sup>1</sup> Anhui Province Key Laboratory of Simulation and Design for Electronic Information System, Hefei Normal University, Hefei, China and <sup>2</sup>School of Computer Science and Technology, Hefei Normal University, Hefei, China

# **Abstract**

The electromagnetic scattering problem over a wide incident angle can be rapidly solved by introducing the compressive sensing theory into the method of moments, whose main computational complexity is comprised of two parts: a few calculations of matrix equations and the recovery of original induced currents. To further improve the method, a novel construction scheme of measurement matrix is proposed in this paper. With the help of the measurement matrix, one can obtain a sparse sensing matrix, and consequently the computational cost for recovery can be reduced by at least half. The scheme is described in detail, and the analysis of computational complexity and numerical experiments are provided to demonstrate the effectiveness.

## **Introduction**

Method of moments (MoM) possesses the advantage of high accuracy in solving the electromagnetic (EM) scattering problems [\[1\]](#page-4-0). However, it will cost a huge computational amount when the incident wave is from a wide-angle range, since the procedure needs to be repeatedly implemented at every angle increment. To accelerate the solution process, many methods have been proposed, such as asymptotic waveform evaluation [\[2\]](#page-4-0), model-based parameter estimation [\[3\]](#page-4-0), etc., but these methods show some shortcomings [\[4\]](#page-4-0). Recently, a fast method based on MoM conjunction with compressive sensing (CS) theory has been put forward [\[5\]](#page-5-0). In this method, a kind of new excitation sources containing abundant information of different incident angles is built firstly.Then, the induced currents over a wide-angle can be solved by means of the sparse transform and the recovery algorithm from the measurement results, which are obtained by a few calculations of traditional MoM with the new sources. The computational complexity of the fast method mainly consists of two parts: one is the measurement, i.e., the calculations of MoM; the other is to acquire the projections of induced currents in sparse domain.

In order to further improve the fast method, much effort has been devoted to the research on these two parts, and many effective schemes, such as Bayesian CS method [\[4\]](#page-4-0), efficient basis function [\[6\]](#page-5-0), and two-dimensional CS method [\[7\]](#page-5-0), have been devised. In this paper, a novel scheme for designing the measurement matrix is raised, by which the sensing matrix shows remarkable sparsity when the orthogonal basis is taken as the sparse transform. Accordingly, the computational complexity for acquire the projections by using recovery algorithm can be sharply decreased.The principle and the complexity analysis are presented, and the effectiveness is validated by the numerical experiments, in which several typical orthogonal bases, such as the fast Fourier transform (FFT) basis, are taken as the sparse transforms, respectively.

# **Theory**

# **Fast method based on CS**

The wide-angle EM scattering problem solving by the traditional MoM can be described as a matrix equation with multiple right-hand sides

> $\mathbf{Z}\left[\mathbf{I}_1\mathbf{I}_2\cdots\mathbf{I}_n\right]=\left[\mathbf{V}_1\mathbf{V}_2\cdots\mathbf{V}_n\right]$  $\Big]$ , (1)

in which, **Z** is the impedance matrix,  $V_1$  to  $V_n$  are the excitation vectors at *n* different incident angles, and  $I_1$  to  $I_n$  represent the *n* corresponding induced current vectors.

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<span id="page-1-0"></span>In the fast method, M new excitation vectors based on CS theory are constructed as

$$
\mathbf{V}_{i}^{\text{CS}} = c_{i1}\mathbf{V}_{1} + c_{i2}\mathbf{V}_{2} + \cdots + c_{in}\mathbf{V}_{n} \ \ (i = 1, 2, \cdots, M), \qquad (2)
$$

where  $c_{ij}$  is the element in the measurement matrix  $\Phi$ . In CS theory, a measurement matrix must satisfy the restricted isometry property (RIP) [\[8\]](#page-5-0), which ensures the accurate reconstruction of original signal. In general, Gaussian random matrix is often used as  $\Phi$ .

Substituting equation  $(2)$  to equation  $(1)$ , one can obtain M current vectors under the new excitations by

$$
\mathbf{Z}\left[\mathbf{I}_1^{\mathrm{CS}}\mathbf{I}_2^{\mathrm{CS}}\cdots\mathbf{I}_M^{\mathrm{CS}}\right]=\left[\mathbf{V}_1^{\mathrm{CS}}\mathbf{V}_2^{\mathrm{CS}}\cdots\mathbf{V}_M^{\mathrm{CS}}\right].\tag{3}
$$

Due to the linearity of the problem,  $I_1^{CS}$  to  $I_M^{CS}$  can be written as

$$
\mathbf{I}_{i}^{\text{CS}} = c_{i1}\mathbf{I}_{1} + c_{i2}\mathbf{I}_{2} + \cdots + c_{in}\mathbf{I}_{n} \ \ (i = 1, 2, \cdots, M). \tag{4}
$$

The M current vectors can be regarded as the results of M measurements of the original induced current vectors  $\mathbf{I}_1$  to  $\mathbf{I}_n$ . If the original induced current vectors have a sparse representation, equation (4) can be described as

$$
\Phi \left[\mathbf{I}_1 \mathbf{I}_2 \cdots \mathbf{I}_n\right]^T = \Phi \Psi \left[\alpha_1 \alpha_2 \cdots \alpha_N\right] = \left[I_1^{CS} I_2^{CS} \cdots I_M^{CS}\right]^T, \tag{5}
$$

in which  $\Psi$  is the sparse transform, N is the number of the basis functions, and  $\alpha_1$  to  $\alpha_N$  are the projections of each column of  $[\mathbf{I}_1 \mathbf{I}_2 \dots \mathbf{I}_n]^T$  in sparse domain. In the wide-angle EM scattering problems, FFT basis, Hermite basis, discrete wavelet transform (DWT) [\[9\]](#page-5-0) basis, or other orthogonal bases are often selected as  $\Psi$  [\[7,](#page-5-0) [10\]](#page-5-0).

By utilizing the recovery algorithm (e.g., orthogonal matching pursuit (OMP)  $[11]$ ), the projections can be approximated by

$$
[\hat{\alpha}_1 \hat{\alpha}_2 \cdots \hat{\alpha}_N] = \operatorname{argmin} \left\| [\hat{\alpha}_1 \hat{\alpha}_2 \cdots \hat{\alpha}_N] \right\|_L \text{ s.t. } \Theta \left[ \alpha_1 \alpha_2 \cdots \alpha_N \right]
$$

$$
= \left[ \mathbf{I}_1^{\text{CS}} \mathbf{I}_2^{\text{CS}} \cdots \mathbf{I}_M^{\text{CS}} \right]^T, \tag{6}
$$

where  $\Theta$  is the sensing matrix and  $\Theta = \Phi \Psi$ .

Then, the original induced current vectors are reconstructed by

$$
\left[\hat{\mathbf{I}}_1 \hat{\mathbf{I}}_2 \cdots \hat{\mathbf{I}}_n\right]^T = \Psi \left[\hat{\boldsymbol{\alpha}}_1 \hat{\boldsymbol{\alpha}}_2 \cdots \hat{\boldsymbol{\alpha}}_N\right]. \tag{7}
$$

The computational complexity for solving equation (6) by OMP is  $O(nKMN)$ , where K is the sparsity of original induced current vectors in sparse domain, and the inner products of the columns in sensing matrix and the measurement results are the dominant computational cost.

### **Construction scheme of measurement matrix**

To reduce the complexity of recovery, one can make the sensing matrix sparse to decrease the cost of inner products. For the purpose, a novel construction scheme of measurement matrix is proposed as follows:

First, by randomly extracting  $P$  columns from the sparse transform  $\Psi$ , one can obtain  $\psi_1$  to  $\psi_P$ , where  $\psi$  represents the column of  $\Psi$ .

Afterward, to better satisfy RIP, a linear superposition of these P vectors is implemented as

$$
\varphi_1 = d_{11}\Psi_1 + d_{12}\Psi_2 + \cdots + d_{1P}\Psi_P.
$$
 (8)

Finally, repeating the above two steps  $M$  times and a new measurement matrix  $[\varphi_1 \varphi_2 ... \varphi_M]^T$  is established, in which

$$
\varphi_i = d_{i1} \Psi_1 + d_{i2} \Psi_2 + \dots + d_{iP} \Psi_P \quad (i = 1, 2, \dots, M). \tag{9}
$$

Obviously, when an orthogonal basis is taken as  $\Psi$ , the inner products of  $\varphi_i$  and the  $(n - P)$  columns in  $\Psi$  that are not extracted in the first step are zero, respectively. In other words, there are only P non-zero elements in the *i*th row of  $\Theta$ . A sparse  $\Theta$  with MP



Figure 1. Relationship between the recovery error and the number of measurements in the case of FFT basis.

<span id="page-2-0"></span>Table 1. Computing time of recovery for sphere (unit: s)

Measurement	FFT.	Hermite	<b>DWT</b>
matrix	$(M = 310)$	$(M = 245)$	$(M = 275)$
Gaussian	8273.0	6513.7	6987.1
Proposed	3085.1	2974.2	2990.2
	$(P = 1250)$	$(P = 1550)$	$(P = 1400)$

non-zero elements is obtained by using the proposed measurement matrix, and the operations of inner products in recovery algorithm are significantly accelerated. Accordingly, the computational complexity of solution to [equation \(6\)](#page-1-0) is decreased to  $O(\eta nKMN)$ , where  $\eta$  is the proportion of non-zero elements in  $\Theta$  and  $\eta = P/n$ . Generally,  $P$  is much less than  $n$ .

Using the proposed measurement matrix and an orthogonal basis as  $\Phi$  and  $\Psi$  respectively, [equation \(5\)](#page-1-0) can be transformed as

$$
[\varphi_1 \varphi_2 \cdots \varphi_M]^T \Psi [\alpha_1 \alpha_2 \cdots \alpha_N] = \mathbf{S} \Psi^T \Psi [\alpha_1 \alpha_2 \cdots \alpha_N]
$$
  
=  $\mathbf{S} [\alpha_1 \alpha_2 \cdots \alpha_N] = [\mathbf{I}_1^{\text{CS}} \mathbf{I}_2^{\text{CS}} \cdots \mathbf{I}_M^{\text{CS}}]^T,$  (10)

in which, the proposed measurement matrix is expressed as the multiplication of a sparse random matrix **S** and the transposed sparse transform  $\Psi$ . The *i*th row of **S** has *P* random coefficients  $d_{i1}$  to  $d_{iP}$  in the columns corresponding to the randomly extracted columns from  $\Psi$  in the *i*th measurement, and other entries in **S** are all zero.

So, equation (10) can be considered as measuring the sparse signals  $\alpha_1$  to  $\alpha_N$  directly with the sparse random matrix **S**, which has been proved to satisfy a different form of RIP, so-called  $RIP(p)$ for  $p$  equal (or very close) to 1 [\[12,](#page-5-0) [13\]](#page-5-0). Therefore, [equation \(6\)](#page-1-0) can produce an accurate solution with high probability by using the proposed measurement matrix. It is interesting to note that, if the random coefficients are all set to 1 in (10), the sparse random matrix is simplified to a sparse binary matrix (SBM) [\[14\]](#page-5-0), which consists of only 0 and 1. SBM is often applied as the measurement matrix in wireless sensor networks, since it is easy to be implemented on hardware and has low complexity [\[15\]](#page-5-0).

#### **Numerical results**

Two numerical experiments with perfect electrical conductor objects of different shapes are presented in this section to validate the proposed scheme, in which the electric field integral equation is established to solve the problems, and OMP is taken as the recovery algorithm. For the convenience of comparison, we define the recovery error as

$$
\Delta = \frac{\left\| \begin{bmatrix} \hat{\mathbf{I}}_1 \hat{\mathbf{I}}_2 \cdots \hat{\mathbf{I}}_n \end{bmatrix} - \begin{bmatrix} \mathbf{I}_1 \mathbf{I}_2 \cdots \mathbf{I}_n \end{bmatrix} \right\|_2}{\left\| \begin{bmatrix} \mathbf{I}_1 \mathbf{I}_2 \cdots \mathbf{I}_n \end{bmatrix} \right\|_2}.
$$
 (11)

#### **Sphere**

A sphere with the radius of 5 m illuminated by the plane waves of 300 MHz is considered, who contains 12,620 Rao-Wilton-Glisson (RWG) basis functions. The waves are set in the xoy plane, and the incident angle is divided into 0.1<sup>∘</sup> , 0.2<sup>∘</sup> , …, 360<sup>∘</sup> .

FFT basis is used as the sparse transform. As is shown in [Fig. 1,](#page-1-0) the similar precision can be achieved at the same number (310) of measurements by applying the Gaussian random matrix and the proposed measurement matrix respectively, while the number of extracting columns P is larger than 1250. It means that one can get a sensing matrix with non-zero elements accounting for 1250/3600  $(\eta)$  in the best case. Thus, the computational cost for acquiring the projections of original induced currents in sparse domain is cut by about two-thirds  $(1 - \eta)$  by using the proposed measurement matrix rather than Gaussian random matrix; meanwhile, the computational complexity of measurement for both is the same. The comparison of the computing time for acquiring the projections is presented in Table 1, which further proves the high efficiency.

To show the universality of the proposed technique for orthogonal bases, FFT basis is, respectively, replaced by Hermite basis and DWT basis. The corresponding results are shown in Figs. 2 and [3.](#page-3-0) It is evident that, for both Hermite basis and DWT basis, much less computational complexity for acquiring the projections is available with the proposed measurement matrix than with Gaussian matrix under the same condition of measurement. The comparisons of



Figure 2. Relationship between the recovery error and the number of measurements in the case of Hermite basis.

<span id="page-3-0"></span>computing time in the case of two bases are also provided in [Table 1.](#page-2-0)

## **Missile model**

Then, consider a missile model who contains 1963 RWG basis functions, and the angle increment is set as 1<sup>∘</sup> . The second kind of Chebyshev basis and Legendre basis is chosen as the sparse transforms, respectively. Other experimental parameters are the same with the previous one. Both Figs. 4 and [5](#page-4-0) indicate that the proposed scheme is also validity for the complex-shaped objects.

By using the second kind of Chebyshev basis and the proposed measurement matrix ( $M = 90$ ,  $P = 100$ ), and setting the elements less than  $10^{-14}$  in the sensing matrix to zero, there are only 9000 non-zero elements in the sensing matrix, which is consistent with the expected number MP. Hence, the solution to [equation](#page-1-0) [\(6\)](#page-1-0) with the help of OMP is accelerated, which is demonstrated

in [Table 2.](#page-4-0) The bistatic radar cross section (RCS) of the missile model illuminated at a random incident angle (take 77<sup>∘</sup> as an example) is also provided in [Fig. 6,](#page-4-0) which agrees well with the result solved by the traditional MoM. From [Table 2](#page-4-0) and [Fig. 6,](#page-4-0) we can see clearly that the computing time can be significantly reduced while the high accuracy is kept by using the proposed measurement matrix.

#### **Conclusion**

A novel scheme for constructing the measurement matrix has been developed. One can get a sparse sensing matrix by adopting the proposed measurement matrix in the solution to wide-angle EM scattering problem based on MoM conjunction with CS. In addition, the number of measurements required for both Gaussian random matrix and the proposed one is the same. Consequently, the computational complexity for acquiring the projections by



Figure 3. Relationship between the recovery error and the number of measurements in the case of DWT basis.



Figure 4. Relationship between the recovery error and the number of measurements in the case of the second kind of Chebyshev basis.

<span id="page-4-0"></span>

Figure 5. Relationship between the recovery error and the number of measurements in the case of Legendre basis.



Figure 6. Comparison of RCS for the missile model illuminated at a random incident angle.

**Table 2.** Computing time of recovery for missile model (unit: s)

Measurement matrix	2nd Chebyshev $(M = 90)$	Legendre $(M = 100)$
Gaussian	120.8	181.2
Proposed	38.7 $(P = 90)$	59.9 $(P = 110)$

using recovery algorithm can be significantly reduced under the condition that the computational cost for measurement remains unchanged.

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**Competing interests.** The authors report no conflict of interest.

#### **References**

- 1. **See KY, Chua EK and Liu Z** (2009) Accurate and efficient evaluation of MoM matrix based on a generalized analytical approach. Progress in Electromagnetics Research **94**, 367–382.
- 2. **Erdemli YE, Gong J, Reddy CJ and Volakis JL** (1998) Fast RCS pattern fill using AWE technique. IEEE Transactions on Antennas and Propagation **46**(11), 1752–1753.
- 3. **Miller EK** (1998) Model-based parameter estimation in electromagnetic III: Applications to EM integral equations. IEEE Antennas and Propagation Magazine **40**(3), 49–66.
- 4. **Zhang H, Zhao X, Lin Z and Sha W** (2016) Fast monostatic scattering analysis based on Bayesian compressive sensing. Applied Computational Electromagnetics Society Journal **31**(11), 1279–1285.
- <span id="page-5-0"></span>5. **Chen MS, Liu FL, Du HM and Wu XL** (2011) Compressive sensing for fast analysis of wide-angle monostatic scattering problems. IEEE Antennas and Wireless Propagation Letters **10**, 1243–1246.
- 6. **Chai SR and Guo LX** (2016) Compressive sensing for monostatic scattering from 3-D NURBS geometries. IEEE Transactions on Antennas and Propagation **64**(8), 3545–3553.
- 7. **Liu Y, Qi Q, Cao XY, Chen MS, Deng GQ, Huang ZX and Wu XL** (2021) Application of two-dimensional compressive sensing to wavelet method of moments for fast analysis of wide-angle electromagnetic scattering problems. International Journal of Antennas and Propagation **2021**,  $1 - 9$
- 8. **Candes EJ, Romberg J and Tao T** (2006) Robust uncertainty principles: Exact signal reconstruction from highly incomplete frequency information. IEEE Transactions on Information Theory **52**(2), 489–509.
- 9. **Chen MS, Sha W and Wu XL** (2009) Solution of arbitrarily dimensional matrix equation in computational electromagnetics by fast lifting waveletlike transform. International Journal for Numerical Methods in Engineering **80**(8), 1124–1142.
- 10. **Cao XY, Chen MS and Wu XL** (2012) Sparse transform matrices and their application in the calculation of electromagnetic scattering problems. Chinese Physics Letters **30**(2), 028401.
- 11. **Yang M and Hoog de F** (2015) Orthogonal matching pursuit with thresholding and its application in compressive sensing. IEEE Transactions on Signal Processing **63**(20), 5479–5486.
- 12. **Zhang B, Liu Y and Wang K** (2014) Restricted isometry property analysis for sparse random matrices. Journal of Electronics & Information Technology **36**(1), 169–174.
- 13. **Gilbert A and Indyk P** (2010) Sparse recovery using sparse matrices. Proceedings of the IEEE **98**(6), 937–947.
- 14. **Berinde R, Gilbert AC, Indyk P, Karloff H and Strauss MJ** (2008) Combining geometry and combinatorics: A unified approach to sparse signal recovery. In 46th Annual Allerton Conference, Illinois, USA.
- 15. **Mamaghanian H, Khaled N, Atienza D and Vandergheynst P** (2011) Compressed sensing for real-time energy-efficient ECG compression on wireless body sensor nodes. IEEE Transactions on Biomedical Engineering **58**(9), 2456–2466.



**Qi Qi** received the M.S. and Ph.D. degrees in electromagnetic field and microwave technology from Anhui University, Hefei, China, in 2013 and 2020, respectively. He is currently an associate professor with the Anhui Province Key Laboratory of Simulation and Design for Electronic Information System, Hefei Normal University. His current research interests include computational electromagnetics and radar signal processing.



**Xinyuan Cao** received the B.S. degree in computer science and technology from the Anhui University of Technology (AHUT), in 2005, the M.S. degree in software engineering from the University of Science and Technology of China (USTC), in 2008, and the Ph.D. degree electromagnetic field and microwave technology from Anhui University, China, in 2013. He is currently a professor with the Anhui Province Key Laboratory of Simulation and Design for Electronic Information System, Hefei Normal University.



**Yi Liu** received the M.S. degree in electromagnetic field and microwave technology from Anhui University (AHU), China, in 2013. Since 2013, she has been with Hefei Normal University (HFNU), China.





in electromagnetic field and microwave technology from Anhui University, Hefei, China, in 2009 and 2016, respectively. He is currently a professor with the Anhui Province Key Laboratory of Simulation and Design for Electronic Information System, Hefei Normal University.

**Meng Kong** received the M.S. and Ph.D. degrees

**Xiaojing Kuang** received the Ph.D. degree in electromagnetic field and microwave technology from Anhui University, Hefei, China, in 2020. She is currently a professor with the Anhui Province Key Laboratory of Simulation and Design for Electronic Information System, Hefei Normal University.



**Mingsheng Chen** received the M.S. and Ph.D. degrees in electromagnetic field and microwave technology from Anhui University, Hefei, China, in 2003 and 2008, respectively. His research interests include fast numerical methods in computational electromagnetics, compressive sensing, and electromagnetic scattering and radiation.