



A CPW-fed dual band four-port MIMO antenna based on liquid crystal polymer for flexible IoT applications

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Research Paper

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Abstract

A 2×2 dual band MIMO flexible antenna with a low profile and CPW feeding is designed for Internet of Things (IoT) domains, operating in 5G (3.3–3.6/4.8–5.0 GHz), WiMAX (5.25–5.85 GHz) and WLAN (5.15–5.35/5.725–5.825 GHz) bands. Two resonant frequencies are generated by L-shaped branches and a rectangular monopole. The mentioned MIMO antenna comprises four elements arranged vertically. With the addition of an orthogonal branch, significant isolation is attained, which exceeds 21 dB in the entire bands. The four-element MIMO antenna is fabricated and experimented to analyze the performance. According to the measurement, the antenna can operate bandwidth covering 3.156–3.84 GHz (19.55%) and 4.638–6.348 GHz (31.13%). Furthermore, the provided four-element MIMO antenna possesses a number of advantageous properties, such as ECC, DG, and TARC, indicating its suitability for 5G/WiMAX/WLAN applications. In accordance with the results of bending measurement and human body influence, it is evident that the presented liquid crystal polymer antenna could be an ideal candidate for integrating into wearable 5G/WiMAX/WLAN devices.

Introduction

The Internet of Things (IoT) has become a hot research area nowadays [1–3], due to the rapid growth of modern communication technologies. As more smart devices require Internet access, the IoT is gaining popularity to meet the needs of people's convenience life. The IoT system is vital for many applications such as intelligent home, intelligent healthcare, intelligent cities and intelligent grid. Based on 5G technology, massive devices in small sizes will be connected to the Internet by IoT systems. MIMO antenna systems can optimize the utilization of space resources by enhancing spectrum efficiency and minimizing co-channel interference [4], which makes them ideal for IoT applications.

Due to their small size, light weight, and bending capacity, flexible antennas have been widely used in wireless body area network IoT applications and other domains [5–8]. There have been a number of proposals for MIMO antennas utilizing flexible materials [9–17]. In [9], a UWB MIMO antenna is demonstrated integrating with DGS based on the jeans working in UWB and WLAN with excellent isolation of 32 dB. In [10], a MIMO antenna based on felt with an overall size of $38.1 \times 38.1 \times 2 \text{ mm}^3$ covering 2.4–2.65 GHz. A MIMO antenna fed by coaxial has been designed using jeans as substrate [11]. The above MIMO antenna configurations with two ports are not suitable for handling additional users. Antennas in [12–17] have great performance with four elements. A flexible substrate has been used in [12, 13] for designing the antenna, which is capable of operating when folded. The antenna in [14] can work in 2.21–6 GHz, which is based on a flexible AgHT-4 substrate. In [15, 16], four-element MIMO antennas using jeans with great isolation have been proposed by Sourav Roy and his team. In [17], a four-port MIMO antenna has been printed on a flexible substrate for 5G NR, Wi-Fi 6, sub-1 GHz and sub-6 GHz applications, but has poor isolation of 10 dB.

The paper proposes a flexible CPW-fed four-element antenna that is fabricated by using a liquid crystal polymer (LCP) substrate and tested in a microwave anechoic chamber. LCP has the properties of thermal stability, bendability, low moisture absorbency and stable dielectric constant, which makes the presented antenna an ideal candidate for IoT applications. The oversize is $60 \times 60 \times 0.1 \text{ mm}^3$. To achieve high port isolation, an orthogonal branch is added in the antenna's center, which connects the ground of different elements. The actual measurement indicates that the operating bands of the four-element MIMO antenna are 3.156–3.84/4.638–6.348 GHz with great isolation of -21 dB , meeting the many spectrum requirements such as 5G, WLAN, WiMAX and so on. Additionally, the antenna performs well in bowed conditions and on the body of the user.

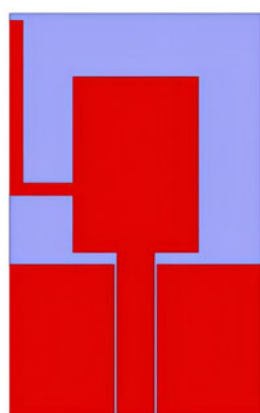
Antenna design

Single element antenna

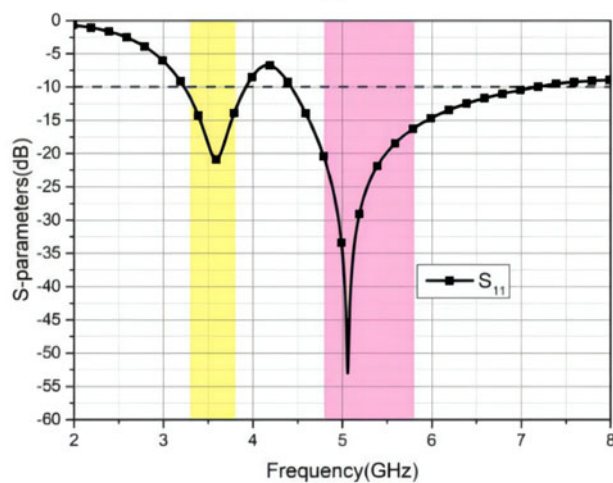
Figure 1 describes the initial geometry of the single antenna element and its S-parameter, the L-shaped branch of the antenna can resonate at 3.5 GHz, while the rectangular patch of the antenna can resonate at 5.5 GHz. The S_{11} (< -10 dB) range of the antenna is 3.23 to 3.92 GHz and 4.43 to 7.17 GHz, which can meet the required frequency range of 3.5 and 5.5 GHz.

Four-element MIMO antenna

An illustration of the geometry of the provided MIMO antenna is exhibited in Fig. 2. It is fed by a coplanar waveguide and printed on a 0.1 mm thick LCP substrate which has a relative dielectric constant = 2.9 and loss tangent = 0.002. It comprises four vertically placed radiation elements and an orthogonal branch is located at the center of the four units to reduce unnecessary coupling, which connects the ground between them. The two resonate points are realized by radiation units comprising a rectangular monopole and an L-shaped branch. All elements are fed by CPW. The following are the final values of simulation and optimization of the different geometric parameters using HFSS:



(a)



(b)

Fig. 1. The structure and S-parameter of single antenna. (a) Single antenna structure (b) S_{11} of the single antenna.

$W = 60$ mm, $W_1 = 10$ mm, $W_2 = 5$ mm, $W_d = 8.3$ mm, $L = 60$ mm, $L_1 = 14$ mm, $L_2 = 13$ mm, $L_3 = 50$ mm.

Isolation structure analysis

A diagram of the decoupling structure design process is exhibited in Fig. 3. Ant-1 is comprised of four elements vertically arranged to each other without isolation branches. Ant-2 adds an orthogonal branch placed in the center which connects the ground of different elements based on Ant-1.

Figure 4 depicts the simulated scattering parameters of both Ant-1 and Ant-2. Because the placement of four elements is centrosymmetric, the verdict of $S_{11} = S_{22} = S_{33} = S_{44}$, $S_{12} = S_{14} = S_{23} = S_{34} = S_{43} = S_{32} = S_{41} = S_{21}$ and $S_{13} = S_{24} = S_{42} = S_{31}$ can be obtained. Due to orthogonality, the perpendicular arrangement of the antenna elements has degraded the coupling between adjacent elements. However, S_{31} is higher than -15 dB in the low-frequency part. The addition of the isolation branch significantly has decreased the values of S_{31} , which means the coupling of two diagonally placed elements has reduced. Even though isolation branches cause coupling between orthogonal antenna elements, S_{21} still meets working requirements. By adding the isolation branch, the operating band ($S_{11} < -10$ dB) covers 3.05–3.71 and 3.89–6.72 GHz, while the values of S_{21} and S_{31} are all less than -21 dB. At the same time, the branch connects the ground of different antenna units so that each port can obtain the same voltage. According to Fig. 5, the distinction of current distribution between Ant-1 and Ant-2 at different frequencies indicates that the mutual coupling current decreases significantly after adding the isolation branch.

Experimental verification

S-parameter

The introduced four-element flexible antenna was manufactured and tested. In Fig. 6, both simulated and tested S-parameter curves are compared. The antenna operated in 3.156–3.84 GHz (19.55%) and 4.638–6.348 GHz (31.13%), covering 5G, WLAN and WiMAX. Figure 6(b) manifests the isolation measured on the operating frequency band is less than -21 dB. Compared to the simulated results, the actual measurement follows a similar trend. In addition, slight variations in curves can be attributed to fabrication errors, the loss of antenna test connector, and a change in the shape of the substrate.

Radiation patterns

As depicted in Fig. 7, the radiation patterns were simulated and measured at 3.5 and 5.5 GHz, in E-plane and H-plane respectively. Taking into account the symmetrical design, only the case where Port 1 is used as the excitation source and the other Ports as load has been tested. The simulated and measured co-polarization patterns are normalized. The 2-D patterns exhibit that both resonance points emit bidirectional radiation in E-plane while emitting omnidirectional radiation in H-plane. At these frequencies, the cross-polarization is generally lower than the co-polarization except for a few degrees. The measurement results are distorted because there is large interference in the 2-D Microwave Anechoic Chamber. At the same time, no power amplifier is used for the antenna, which also leads to a certain deviation between the simulated pattern and the measured one.

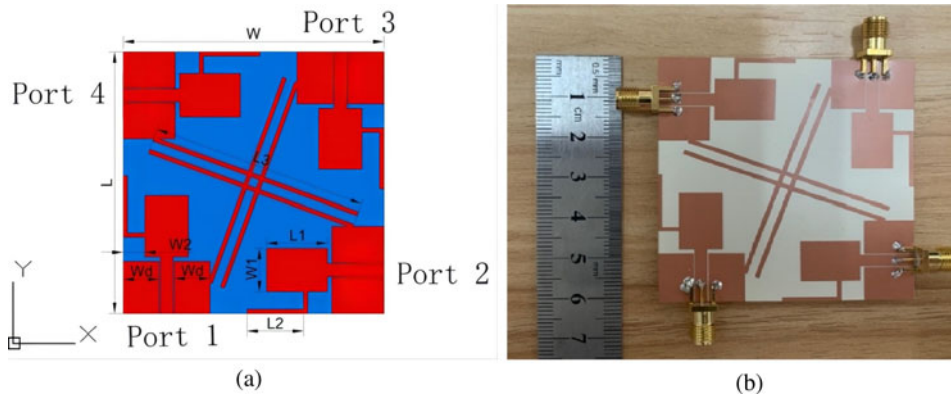


Fig. 2. The introduced MIMO antenna. (a) Geometrical details (b) Fabricated antenna.

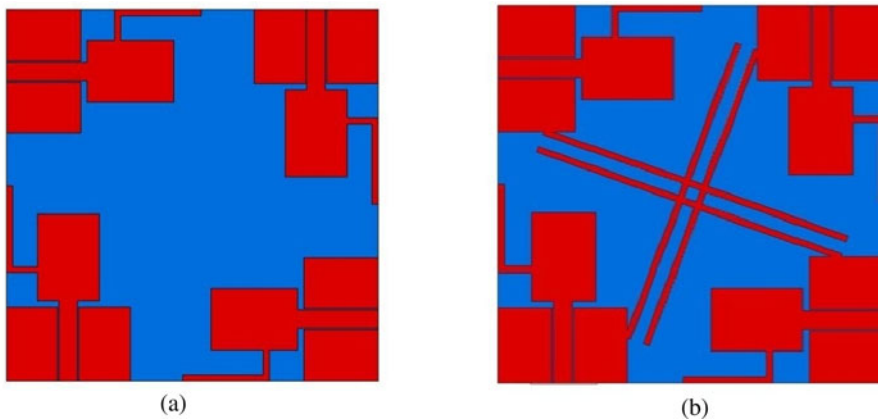


Fig. 3. Design process. (a) Ant-1 (b) Ant-2.

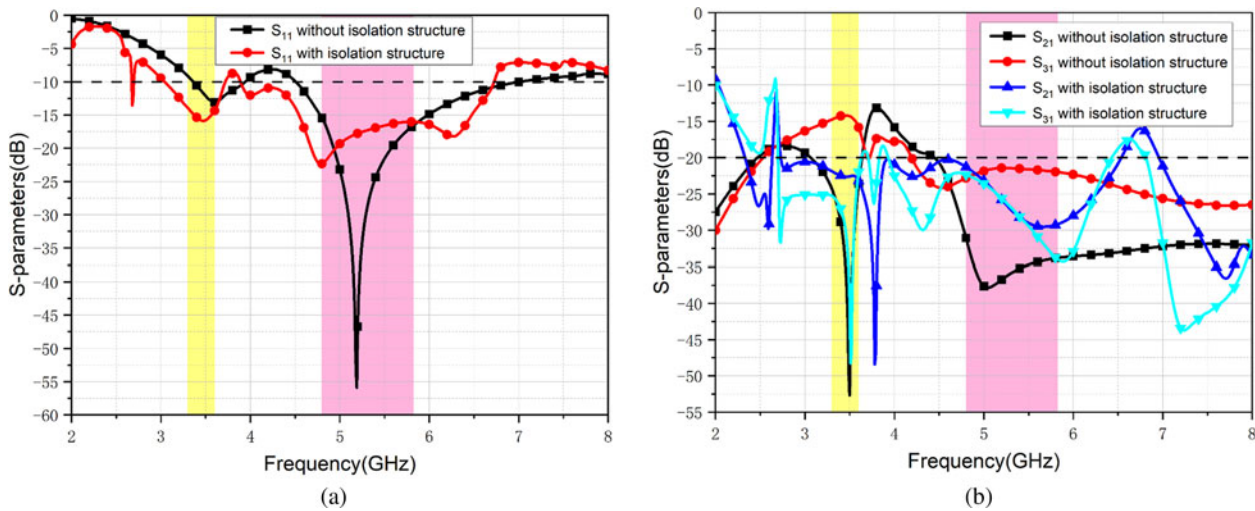


Fig. 4. Simulated S-parameters. (a) S_{11} (b) S_{21} and S_{31} .

ECC and DG of antenna

As a basic parameter, envelope correlation coefficient (ECC) is used in MIMO system to estimate the correlation degree between different antenna elements, which can be calculated as follows [18]:

$$\rho_e(i, j, N) = \left| \frac{\sum_{n=1}^N S_{i,n}^* S_{n,j}}{\prod_{k=i,j} \left(1 - \sum_{n=1}^N S_{k,n}^* S_{n,k} \right)^{1/2}} \right|^2 \quad (1)$$

The antenna must have a relatively high efficiency to maintain the validity of equation (1). By using the following equation (2), then, we can also calculate the ECC more accurately based on the far-field radiation:

$$\rho_e(i, j) = \frac{\left| \iint_{4\pi} \vec{F}_i(\theta, \phi) \cdot \vec{F}_j(\theta, \phi) d\Omega \right|^2}{\iint_{4\pi} \left| \vec{F}_i(\theta, \phi) \right|^2 d\Omega \cdot \iint_{4\pi} \left| \vec{F}_j(\theta, \phi) \right|^2 d\Omega} \quad (2)$$

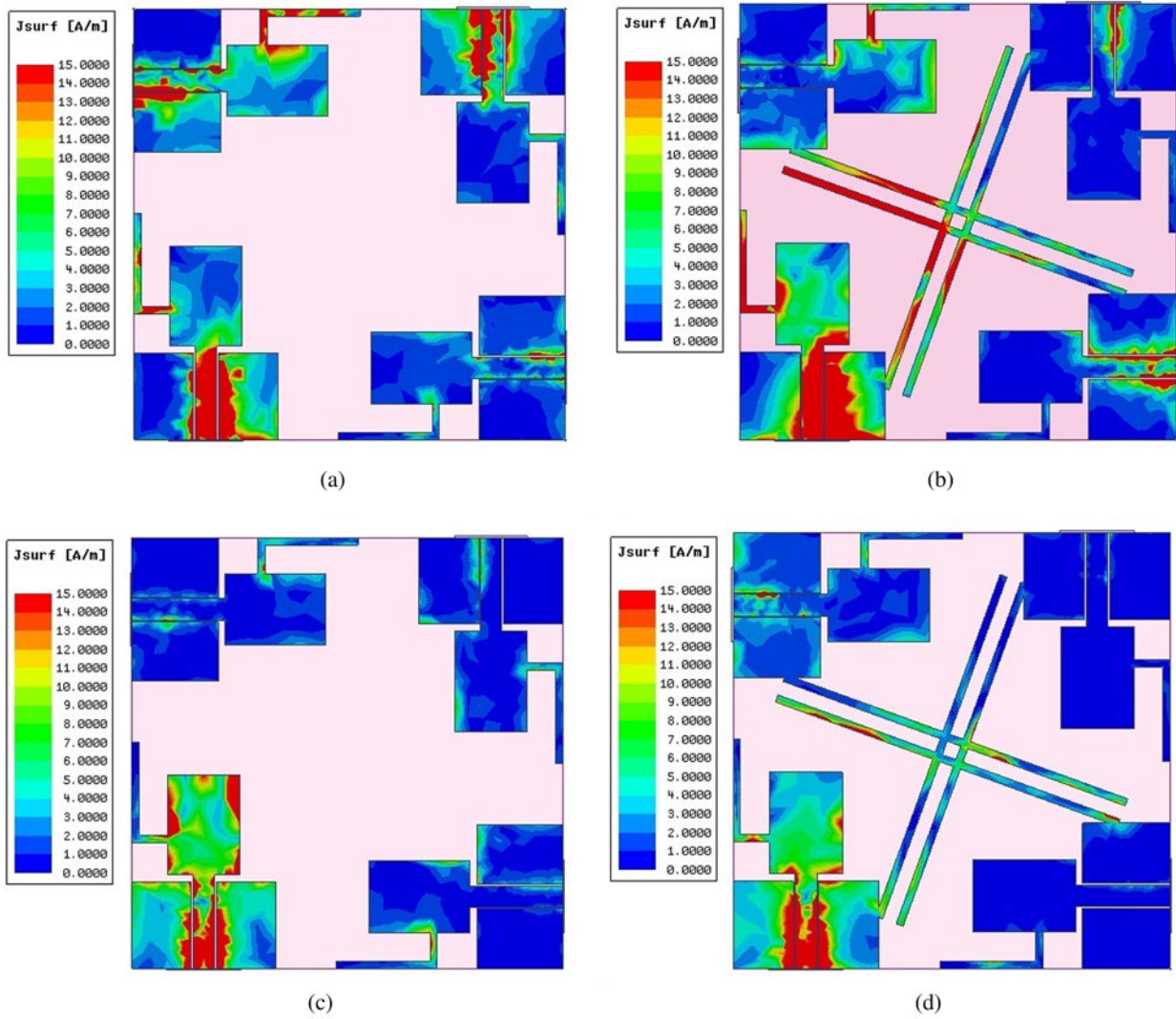


Fig. 5. Current distributions. (a) Ant-1 at 3.5 GHz (b) Ant-2 at 3.5 GHz (c) Ant-1 at 5.5 GHz (d) Ant-2 at 5.5 GHz.

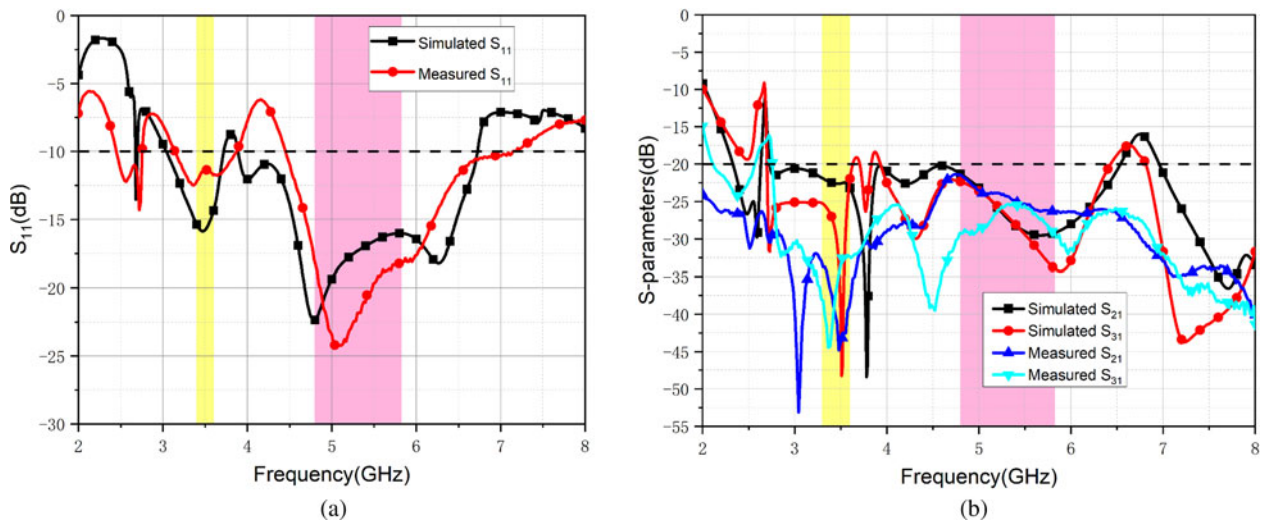


Fig. 6. Simulated and measured S-parameters. (a) S_{11} (b) S_{21} and S_{31} .

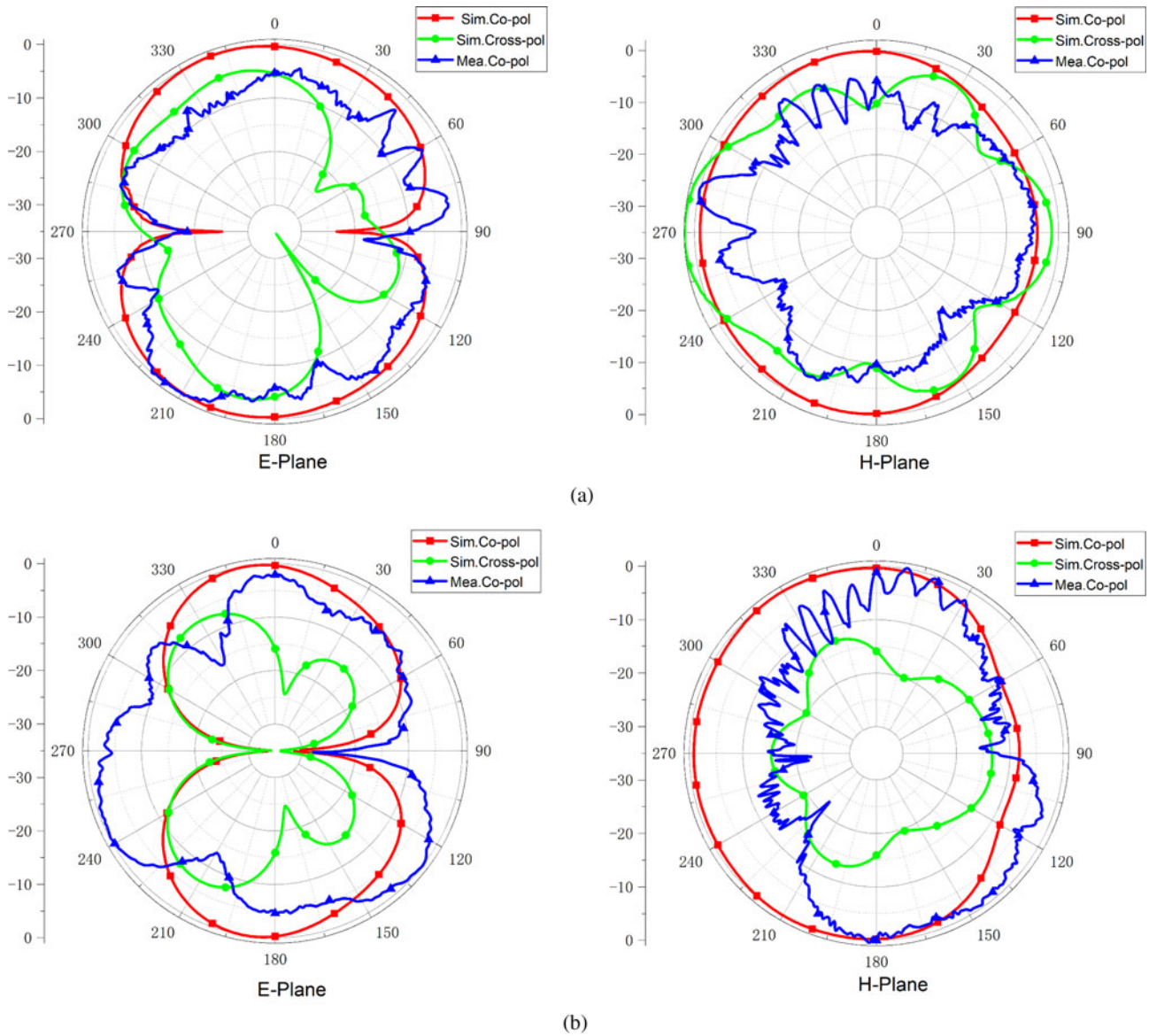


Fig. 7. Radiation patterns. (a) 3.5 GHz (b) 5.5 GHz.

At the same time, diversity gain (DG) is also crucial when evaluating MIMO antenna characteristics. The DG can be calculated as [19]:

$$DG = 10\sqrt{1 - ECC^2} \tag{3}$$

Generally, in the working band, the ECC value should not exceed 0.5, while the DG is larger than 9.5, which can be used in practice. Figure 8 demonstrates that the presented antenna possesses a good ECC less than 0.12, exceeding the requirement by a wide margin.

In Fig. 9, DG curves during 2–8 GHz show a high value of DG, better than 9.93.

Gain

Figure 10 shows the peak gain of simulation and measurement covering 3.0–6.0 GHz. In these results, the measured peak gain is about 4 dB for 3.5 GHz and 4.8 dB for 5.5 GHz. Besides, the gain of measurement displays a lower level compared with that

of simulation, due to the uncertainty of the environment and the difference of measurement.

TARC (total active reflection coefficient)

TARC is another vital indicator, predicting the performance of MIMO system, which can be expressed as [20]:

$$TARC = N^{-1/2} \sqrt{\sum_{i=1}^N \left| \sum_{k=1}^N S_{ik} e^{j\theta_{k-1}} \right|^2} \tag{4}$$

In this expression, θ stands for the excited phase angle. Seven phase angles from 0° to 180° are picked to analyze the properties of the proposed four-element antenna. Figure 11 depicts the values of measured TARC between 2.0 to 8.0 GHz. The measured TARC is less than -10 dB over the functioning bands, indicating that the MIMO antenna has stable characteristics and allowable performance.

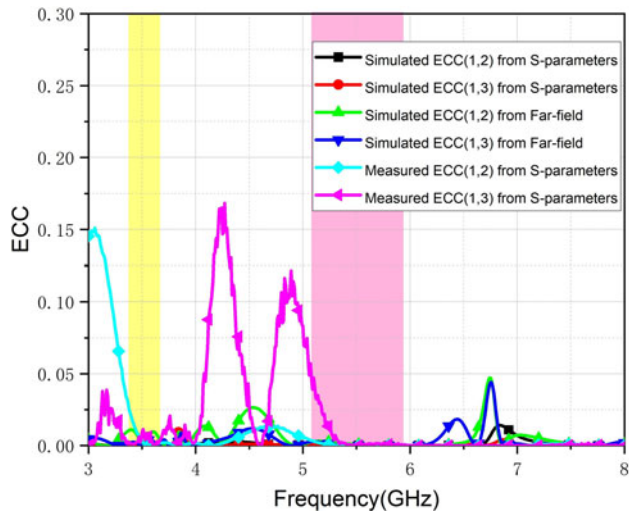


Fig. 8. ECC of the introduced antenna.

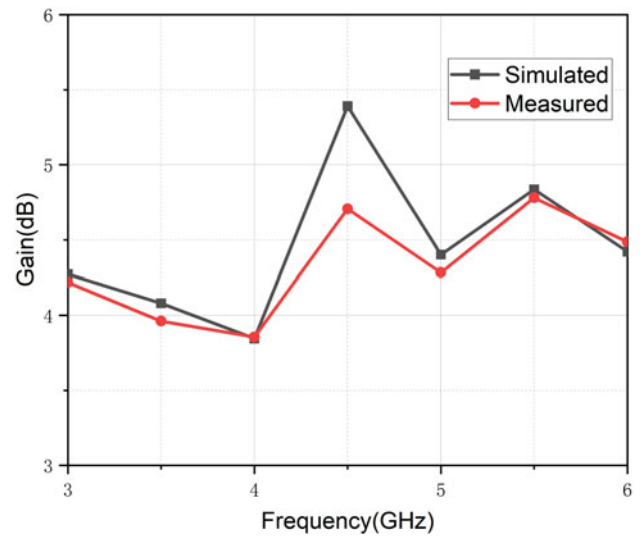


Fig. 10. Simulated and measured gain.

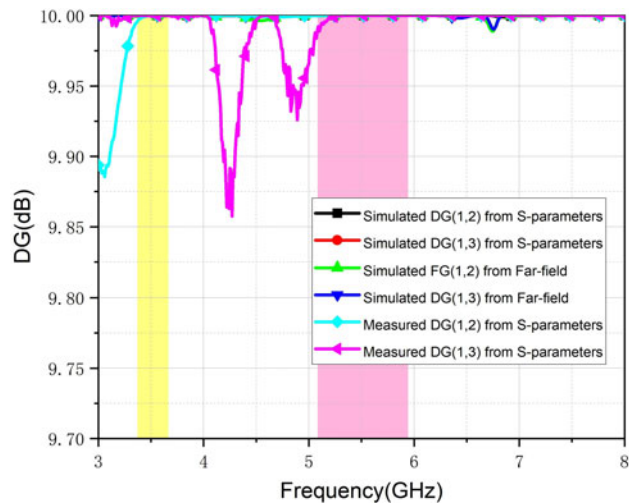


Fig. 9. DG of the introduced antenna.

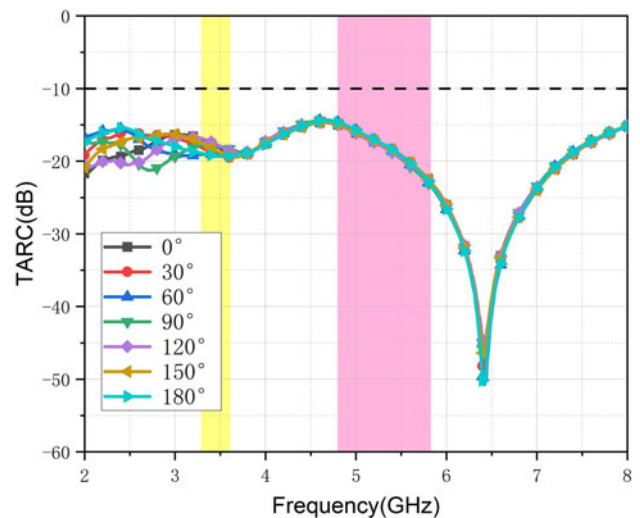


Fig. 11. Measured curves of TARC in different degrees.

Flexibility analysis

The performance of a flexible MIMO antenna under bending conditions is an important metric. The *S*-parameters of this presented four-port flexible antenna are tested when the antenna is fixed on various cylinders with two curvatures ($R = 30$ mm and $R = 45$ mm) along E-plane or H-plane, respectively. Figure 12 depicts the results of return loss and isolation in different conditions. After bending, the antenna retains its original bandwidth. In the low-frequency band, the isolation of measurement is less than -23 dB, while that does not exceed -15 dB in the high-frequency part, which meets the practical standards. The curves of S_{21} , S_{31} and S_{41} show the proposed four-element antenna exhibits good isolation at bowed conditions.

Effects of human body

Since the proposed antenna is a potential candidate for IoT and wearable fields, it is necessary to study the interaction between it and human tissue. The Specific Absorption Rate (SAR) is a

vital parameter towards quantifying the effects of antenna radiation on humans. A human tissue model made up of four layers is designed by HFSS. Figure 13 depicts the model and Table 1 contains its associated parameters.

Parameter *H* represents the distance from the tissue model to an antenna. It is described in Table 2 that maximum SAR values of the presented antenna were attained in various conditions. The simulated values all meet the European Union Standard (2 W/kg/ 10 g).

When the antenna is worn on the arm, chest and thigh, the tested results of scattering parameters are depicted in the following Fig. 14, which indicates that the presented four-element antenna is still in compliance with the working requirements.

Performance comparison

In Table 3, a comparison is shown between the proposed flexible dual band four-element MIMO antenna and previous designs.

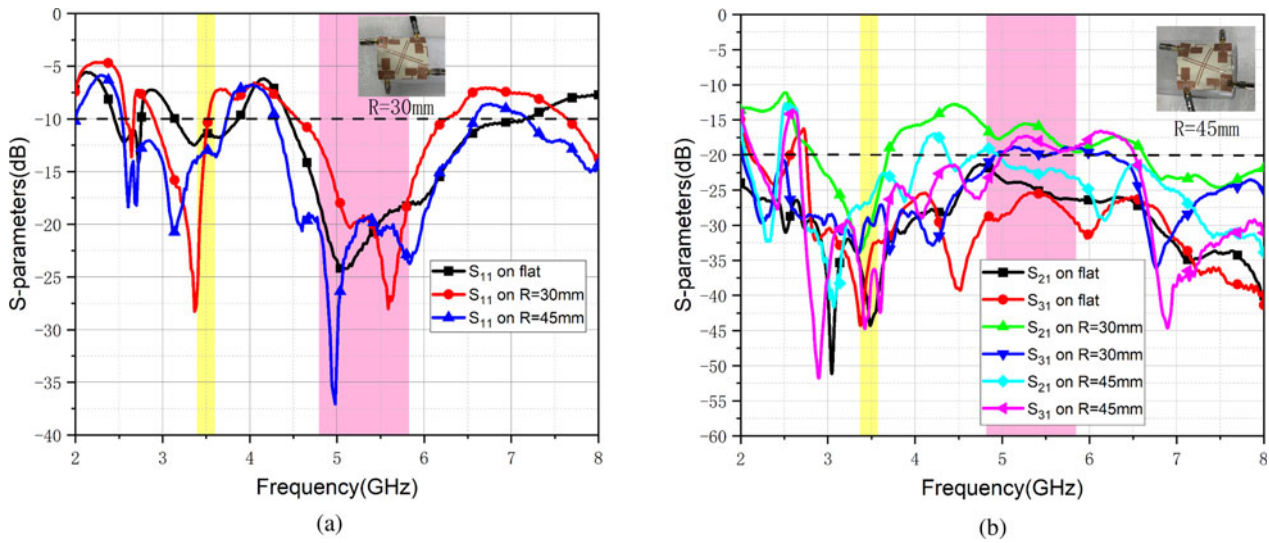


Fig. 12. S-parameters in different bending conditions. (a) S_{11} (b) S_{21} and S_{31} .

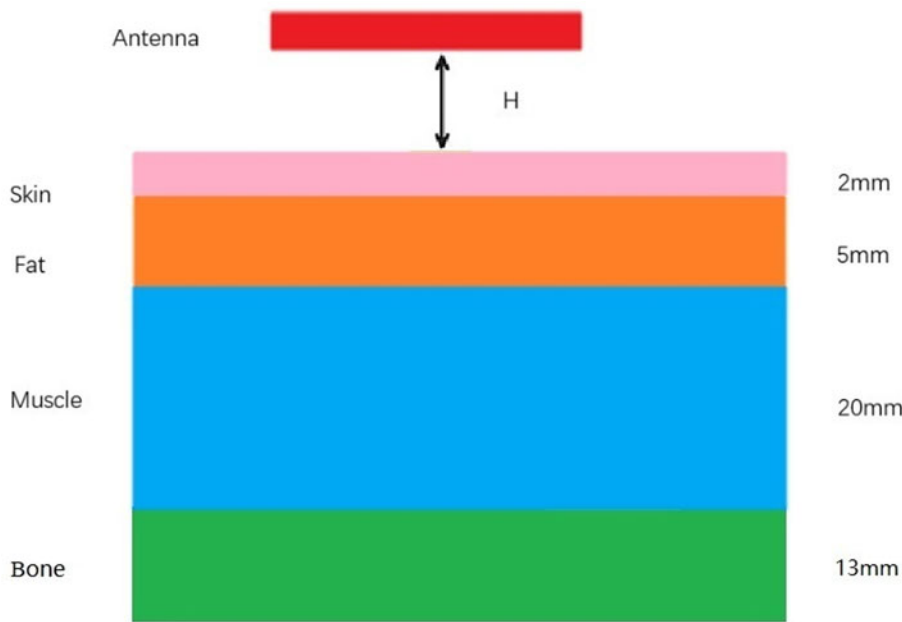


Fig. 13. The human body approaching the antenna model.

Table 1. Characteristics of different tissues of the human body

	Bone	Muscle	Fat	Skin
ϵ_r	18.49	52.67	5.27	37.95
σ (S/m)	0.82	1.77	0.11	1.49
Density (kg/m^3)	1008	1006	900	1001
Thickness (mm)	13	20	5	2

Table 2. Maximum SAR values on various conditions

Frequency (GHz)	H(mm)	SAR (W/Kg/10 g)
3.5	1	0.49
	3	0.13
	5	0.17
5.5	1	1.21
	3	0.69
	5	0.52

The antennas of the literature [10, 11] are only two-port antennas. Thus, their transmission capacity is lower than the one presented. Antennas in [12, 16, 17] are larger than the proposed one. The antenna presented here has better isolation than those in [14,

15]. The table reveals that the introduced flexible MIMO antenna shows a splendid capability with low values of ECC, high isolation and good SAR, which makes it an ideal candidate for IoT devices.

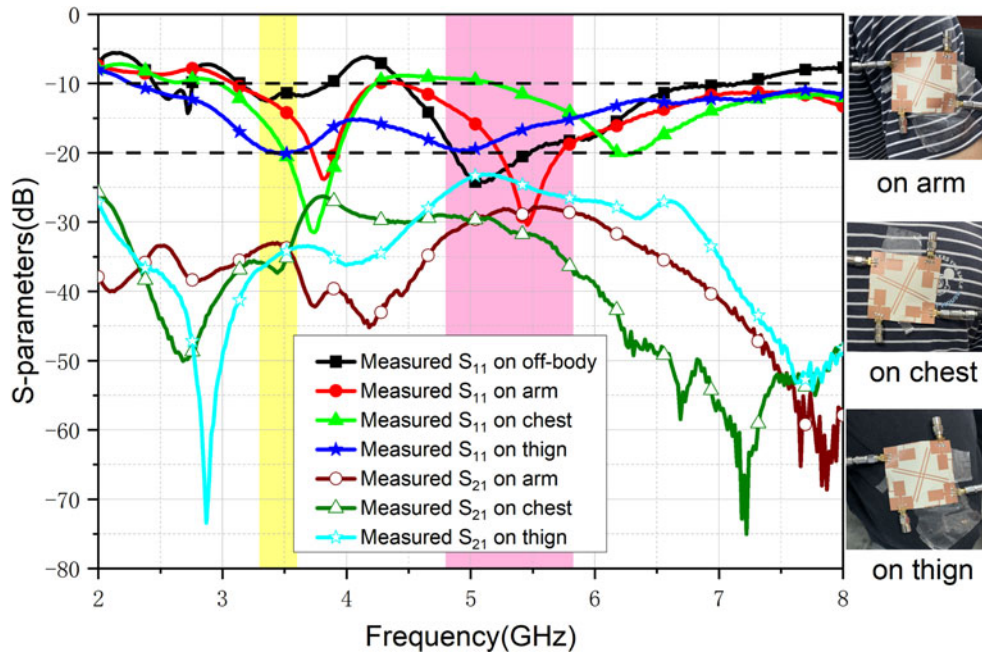


Fig. 14. The tested performance on human body.

Table 3. Comparison between the designed antenna and others

Ref.	No. of ports	Size (mm ³)	Feeding method	Substrate	Bands (GHz)	Iso. (dB)	ECC	DG
[10]	2	38.1 × 38.1 × 2	CPW	Felt	2.4–2.65	10	<0.01	–
[11]	2	101.9 × 92.3 × 3	Coaxial	Felt	2.4–2.5 5.2–5.8	20	–	–
[12]	4	168 × 96 × 1.2	Microstrip	Felt	1.4–5	15	<0.8	–
[14]	4	66 × 45 × 0.625	Microstrip	AgHT-4	2.21–6	15	<0.016	9.9
[15]	4	55 × 58 × 1.4	Microstrip	Jeans	2.74–4.41	19	<0.2	>9.6
[16]	4	110 × 97 × 1.4	Microstrip	Jeans	1.5–3.8 4.1–6.1	24	<0.1	9.9
[17]	4	180 × 180 × 1.52	CPW	Rogers RO3003	0.7–1.01 2.6–3.18 5.3–6.06 6.7–6.94	10	<0.5	–
Prop.	4	60 × 60 × 0.1	CPW	LCP	3.156–3.84 4.638–6.348	21	<0.12	>9.9

Conclusion

A compact CPW-fed flexible dual band four-element MIMO antenna is presented in this work. The operating band is 3.156–3.84 (19.55%)/4.638–6.348 (31.13%), which is compatible with wireless spectrum such as 5G, WLAN, WiMAX, etc. The addition of an orthogonal branch accomplishes decoupling. Therefore, the tested isolation exceeds 31 dB within 3.5 GHz and 21 dB within 5.5 GHz, respectively. Moreover, the flexible four-port antenna exhibits that ECC is less than 0.12, the DG exceeds 9.93, and the TARC curves display excellent characteristics, which indicates it is capable of supporting 5G/WiMAX/WLAN communications.

Owing to the bending resistance and stability, the introduced antenna is expected to function in the wearable technology and IoT domains.

Data. Data sharing not applicable to this article as no datasets were generated or analyzed during the current study.

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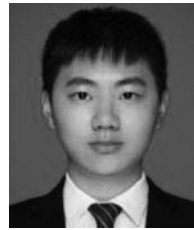
Author contributions. Hongye Liu and Chengzhu Du derived the theory, Jie Zhang and Lingru Pei performed the simulations. All authors contributed equally to analyzing data and reaching conclusions, and in writing the paper.

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Conflict of interest. The authors report no conflict of interest.

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